**Features** 



## **MAX15101**

# **Small 1A, Low-Dropout Linear Regulator** in a 2.7mm x 1.6mm Package

#### **General Description**

♦ 1.7V to 5.5V Operating Range ♦ 100mV Dropout Guaranteed at 1A

The MAX15101 is a small, low-dropout linear regulator optimized for networking, datacom, and server applications. The regulator delivers up to 1A from a 1.7V to 5.5V input supply with output accuracy of ±1.6%. The output is adjustable down to 0.6V with a guaranteed dropout voltage of less than 100mV at full load. The internal p-channel switch is protected with a foldback current limit and thermal shutdown.

- ♦ ±1.6% Output Accuracy Over Line, Load, and **Temperature**
- The MAX15101 features an external bypass input to
- ◆ Adjustable Output from 0.6V to 5.2V

reduce noise. This bypass input also serves as a softstart control. An enable input and power-good output are included for power sequence control.

**♦ Foldback Current Protection** 

The MAX15101 is available in a 1.6mm x 2.7mm WLP package and is fully specified from the -40°C to +85°C operating temperature range.

- ♦ Thermal Shutdown Protection
- ♦ Low Noise 15μV<sub>RMS</sub> (typ)
- ♦ Small 1.6mm x 2.7mm WLP Package

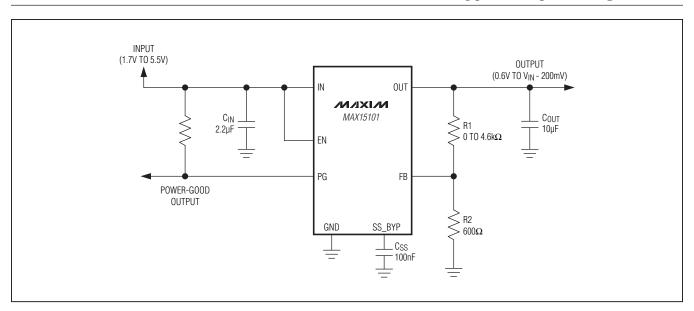
#### **Applications**

Ordering Information appears at end of data sheet.

Networking Telecom/Datacom Servers **Base Stations PLL Power** 

For related parts and recommended products to use with this part, refer to www.maxim-ic.com/MAX15101.related.

### **Typical Operating Circuit**



NIXIN

Maxim Integrated Products 1

# Small 1A, Low-Dropout Linear Regulator in a 2.7mm x 1.6mm Package

#### **ABSOLUTE MAXIMUM RATINGS**

IN, EN, PG, SS BYP, FB to GND0.3V to +6V	Operating Temperature Range40°C to +85°C
IN, EN, FG, 33_BTF, FB to GND0.3V to +0V	Operating reinperature hange40 C to +65 C
OUT to GND0.3V to (V <sub>IN</sub> + 0.3V)	Storage Temperature Range65°C to +150°C
OUT Short Circuit	Junction Temperature+150°C
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	Soldering Temperature (reflow) (Note 1)+260°C
WLP (derate 19.2mW/°C above +70°C) 1538mW	

Note 1: The WLP package is constructed using a unique set of package techniques that impose a limit on the thermal profile. The device can be exposed during board-level solder attach and rework. This limit permits only the use of the solder profiles recommended in the industry-standard specification JEDEC 020A, paragraph 7.6, Table 3 for IR/VPR and convection reflow. Preheating is required. Hand or wave soldering is not allowed.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### PACKAGE THERMAL CHARACTERISTICS (Note 2)

Junction-to-Ambient Thermal Resistance (0,1A) .......30°C/W

Note 2: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to www.maxim-ic.com/thermal-tutorial.

#### **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = 3.3V, V_{OUT} = 1.5V, V_{EN} = V_{IN}, I_{OUT} = 100 \text{mA}, C_{IN} = 2.2 \mu\text{F}, C_{OUT} = 22 \mu\text{F}, T_A = T_J = -40 ^{\circ}\text{C}$  to +85  $^{\circ}\text{C}$ , typical values are at  $T_A = +25$ °C, unless otherwise noted.) (Note 3)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
IN						
Input Voltage Range	V <sub>IN</sub>		1.7		5.5	V
Input Undervoltage Lockout	V <sub>IN_UV</sub>	V <sub>IN</sub> rising		1.624	1.69	V
Input Undervoltage Lockout Hysteresis	V <sub>IN_UV_HYS</sub>			40		mV
OUT						
Output Voltage Range	V <sub>OUT</sub>		0.6		5.2	V
Load Regulation		I <sub>OUT</sub> = 30mA to 1A		0.1		%/A
Line Regulation		$V_{IN} = 1.7V \text{ to } 5.5V, V_{OUT} = 0.6V, I_{OUT} = 190\text{mA}$	-0.15		+0.15	%/V
Dropout Voltage		I <sub>OUT</sub> = 1A, V <sub>FB</sub> = 580mV (Note 4)		25	100	mV
In-Regulation Current Limit		V <sub>IN</sub> = 5.5V, V <sub>FB</sub> = 500mV		1.62		А
Short-Circuit Foldback Current Limit		V <sub>IN</sub> = 5.5V, V <sub>OUT</sub> = 0V, V <sub>FB</sub> < 100mV		1		А
Output Shutdown Leakage		$V_{IN} = 5.5V, V_{OUT} = 0V, V_{EN} = 0V$			4	μΑ
Line Transient		$\Delta V_{IN} = 0.5 V$ , $t_{RISE} = t_{FALL} = 5 \mu s$		2		%

# **Small 1A, Low-Dropout Linear Regulator** in a 2.7mm x 1.6mm Package

#### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{IN}=3.3V,\,V_{OUT}=1.5V,\,V_{EN}=V_{IN},\,I_{OUT}=100$ mA,  $C_{IN}=2.2\mu$ F,  $C_{OUT}=22\mu$ F,  $C_{OUT}=22\mu$ F,  $C_{OUT}=240$ C to +85°C, typical values are at  $C_{IN}=4.25$ °C, unless otherwise noted.) (Note 3)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Load Transient		$I_{OUT}$ = 100mA to 1A to 100mA, $t_{RISE}$ = $t_{FALL}$ = 1 $\mu$ s			1.3		%
Power-Supply Rejection		Frequency = 100kHz, I <sub>OUT</sub>	= 500mA		40		dB
Output Voltage Noise		Frequency = 500Hz to 100kHz, I <sub>OUT</sub> = 100mA			15		μV <sub>RMS</sub>
FB							
FB Threshold Accuracy		$I_{OUT} = 190$ mA, $V_{OUT} = 1.5$ V to 5V, $V_{IN} = (V_{OUT} + 0.3$ V) to 5.5V		593	600	606	m\/
FB Tilleshold Accuracy		$I_{OUT} = 3mA \text{ to } 1A, V_{OUT} = V_{IN} = (V_{OUT} + 0.3V) \text{ to } 5.5V$	$I_{OUT} = 3mA$ to 1A, $V_{OUT} = 1.5V$ to 5V,			610	- mV
FB Input Current		$V_{FB} = 0.6V, V_{IN} = 5.5V$				1	μΑ
GND							
Ground Supply Current					1.8	3.5	mA
Ground Shutdown Current		V <sub>IN</sub> = 1.7V to 5.5V				1	μΑ
POWER-GOOD OUTPUT							
Dawar Coad (DC) Throubald		V <sub>FB</sub> moving out of	FB = high	638	650	662	mV
Power-Good (PG) Threshold		regulation, V <sub>HYS</sub> = 15mV	FB = low	538	550	562	
Power-Good (PG) Output- Voltage Low		I <sub>PG</sub> = 2mA (sink)			25	80	mV
Power-Good (PG) Delay					100		μs
Power-Good (PG) Leakage		V <sub>PG</sub> = 5.5V				1	μΑ
ENABLE							
Enable Input High Threshold	V <sub>EN_HIGH</sub>			1.2			V
Enable Input Low Threshold	V <sub>EN_LOW</sub>					0.4	V
Enable Input Bias Current		V <sub>EN</sub> = 0 to 5.5V		-1		+1	μΑ
SOFT-START							
Soft-Start Charging Current		V <sub>SS_BYP</sub> = 100mV		8	9.2	12	μΑ
Soft-Start Pulldown Resistance		$V_{EN} = 0V$ , $I_{SS\_BYP} = 2mA$			45	100	Ω
THERMAL SHUTDOWN							
Thermal Shutdown Threshold					+160		°C
Thermal Shutdown Hysteresis		Threshold falling			15		°C

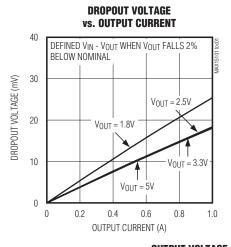
**Note 3:** All devices 100% production tested at  $T_A = +25$ °C. Limits over the temperature range are guaranteed by design.

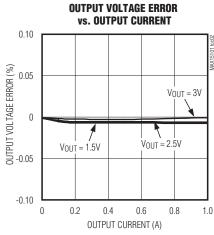
Note 4: Dropout is defined as V<sub>IN</sub> - V<sub>OUT</sub> when V<sub>OUT</sub> falls 3% below its nominal voltage value.

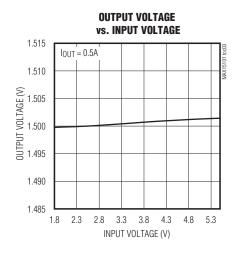
# **Small 1A, Low-Dropout Linear Regulator** in a 2.7mm x 1.6mm Package

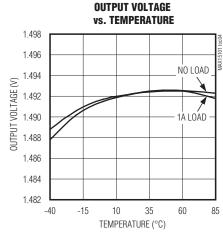
## **Typical Operating Characteristics**

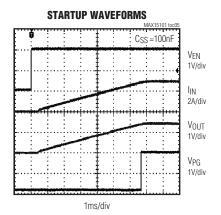
 $(V_{IN} = 1.8V, V_{OUT} = 1.5V, T_A = +25$ °C, unless otherwise noted.)

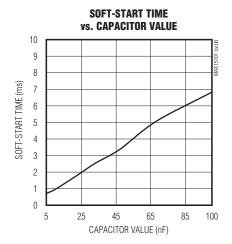


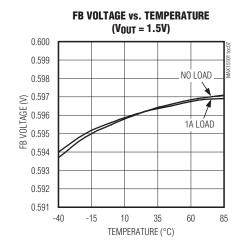








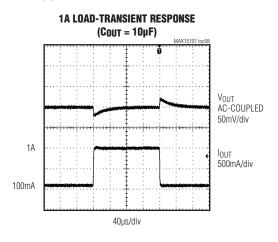


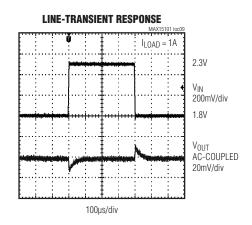


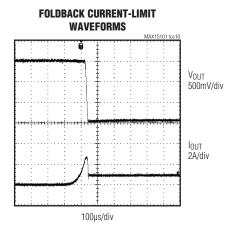
# **Small 1A, Low-Dropout Linear Regulator** in a 2.7mm x 1.6mm Package

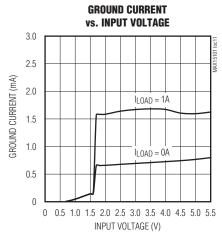
## **Typical Operating Characteristics (continued)**

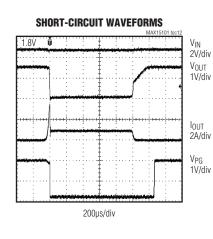
 $(V_{IN} = 1.8V, V_{OUT} = 1.5V, T_A = +25$ °C, unless otherwise noted.)

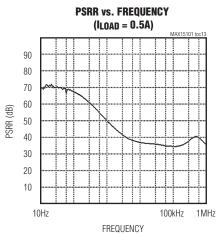


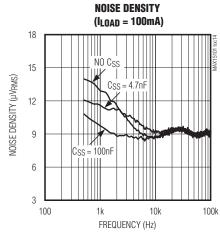






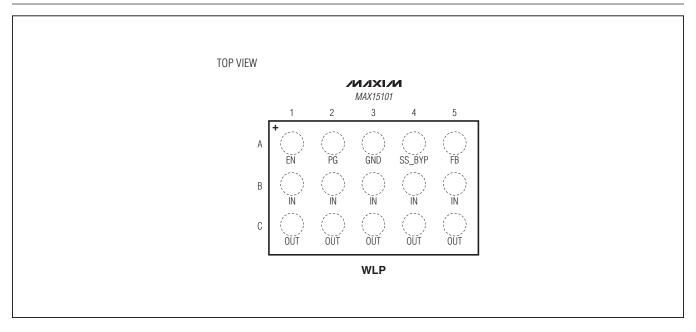






# **Small 1A, Low-Dropout Linear Regulator** in a 2.7mm x 1.6mm Package

## **Bump Configuration**

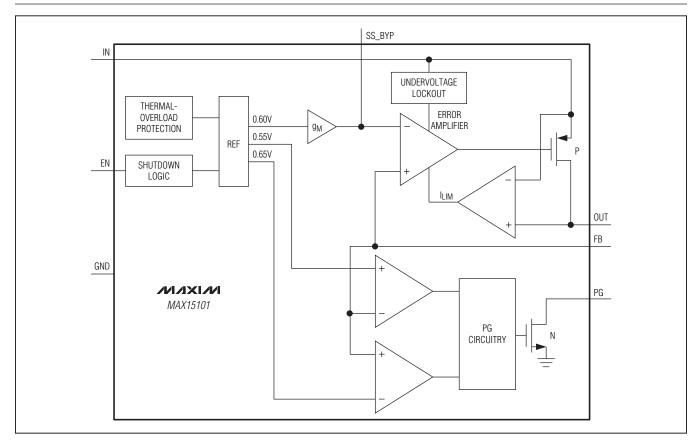


## **Bump Description**

BUMP	NAME	FUNCTION
A1	EN	Active-High Enable Input. Drive EN logic-high to enable the LDO. Connect to IN for always-on operation.
A2	PG	Power-Good Output. Open-drain output asserts high when the voltage is within regulation.
A3	GND	Ground
A4	SS_BYP	SS_BYP Bypass/Soft-Start Input. Connect a ceramic capacitor from SS_BYP to GND to reduce the output noise and set the soft-start ramp time.
A5	FB	Feedback Input. Connect a resistive divider from OUT to GND to set the output voltage.
B1, B2, B3, B4, B5	IN	Supply Voltage Input. Connect to supply voltage from 1.7V to 5.5V. Bypass IN to GND with a 2.2µF or greater ceramic capacitor.
C1, C2, C3, C4, C5	OUT	Regulated Output. Connect a 10µF capacitor from OUT to GND. For better load transient performance a larger capacitor can be used.

# **Small 1A, Low-Dropout Linear Regulator** in a 2.7mm x 1.6mm Package

### Functional Diagram



## **Detailed Description**

The MAX15101 is a small, low-dropout linear regulator optimized for server, network, and telecom applications. The regulator delivers up to 1A from a 1.7V to 5.5V input supply with output accuracy of ±1.6%. The output is adjustable down to 0.6V with a guaranteed dropout voltage of less than 100mV at full load. The internal p-channel switch is protected with a foldback current limit and thermal shutdown.

The MAX15101 features an external bypass input to reduce noise. This bypass input also serves as a softstart control. An enable input and power-good output are included for power sequence control.

#### **Internal p-Channel Pass Transistor**

The MAX15101 features a 1A p-channel MOSFET pass transistor. Unlike similar designs with pnp pass transistors, the gate-drive current to the p-channel power transistor is extremely low. PNP-based regulators also waste considerable current in dropout when the pass transistor saturates and uses high base-drive currents under large loads. The MAX15101 does not suffer from these deficiencies and consumes only 1.8mA of guiescent current even in dropout.

#### **Output Voltage Selection**

The MAX15101 features an adjustable output. The output voltage can be set between +0.6V to VIN - 200mV using an external resistive voltage-divider to FB (see the Typical Operating Circuit). The output voltage is set using the following equation:

$$V_{OUT} = V_{FB} (1 + R1/R2)$$

where  $V_{FB} = +600$ mV.

To simplify resistor selection:

$$R1 = R2 (V_{OUT}/V_{FB} - 1)$$

Set R2 to  $600\Omega$ .

# **Small 1A, Low-Dropout Linear Regulator** in a 2.7mm x 1.6mm Package

#### **Enable**

A logic-low on EN disables the MAX15101. In shutdown, the pass transistor, control circuitry, reference, and all biases are turned off, reducing supply current to typically 1µA. Connect EN to IN for normal operation. In shutdown, the soft-start capacitor is discharged using an internal pulldown resistance.

#### Soft-Start

As shown in the Typical Operating Circuit, the MAX15101 uses a g<sub>M</sub> stage and integrating capacitor C<sub>SS</sub> to control the regulator's feedback set point. When EN is driven low, the soft-start capacitor is discharged. When EN is driven high or power is applied to the device, a constant 9.2µA current charges the soft-start capacitor. At startup, the feedback set point increases linearly, which reduces the inrush supply current.

$$t_{SS} = 6.315 \times 10^{-5} \times C_{SS}$$

where Css is in nF. Use capacitor values between 30nF to 100nF with voltage ratings greater than 5V.

Because this ramp is applied to the current-limit comparator, the actual time for the output voltage to ramp up depends on the load current and output capacitor. Leave SS\_BYP unconnected to disable soft-start.

#### **Foldback Current Limiting**

The MAX15101 features a 1.6A current limit when the output voltage is in regulation. When the voltage at the output drops to 17% of its nominal value, the current limit exponentially folds back to 1A. The output can be shorted to ground for an indefinite period of time without damaging the device. Avoid continuous output current of greater than 1A to prevent damage to the regulator.

#### **Thermal-Overload Protection**

A thermal-overload circuit limits total power dissipation in the MAX15101. When the junction temperature exceeds  $T_{JJ} = +160$ °C, the thermal sensor turns off the pass transistor, allowing the IC to cool. The thermal sensor turns the pass transistor on once the IC's junction temperature drops approximately 15°C. Continuous short-circuit conditions will consequently result in a foldback limited output.

Thermal-overload protection is designed to safeguard the MAX15101 in the event of fault conditions. For continuous operation, do not exceed the junction temperature rating of  $T_J = +125$ °C.

#### **Operating Region and Power Dissipation**

The power dissipation across the device is  $P = I_{OUT}$ (V<sub>IN</sub> - V<sub>OUT</sub>). The resulting maximum power dissipation is:

$$P_{MAX} = [(T_{J(MAX)} - T_{A})/(\theta_{JA})]$$

where  $(T_{J(MAX)}$  -  $T_{A})$  is the temperature difference between the maximum allowed die junction (+150°C) and the surrounding air. Figure 1 shows the allowable power dissipation for typical PCBs at +25°C, +50°C, and +70°C ambient temperatures (refer to the MAX15101 EV kit).

#### **Applications Information**

#### **Capacitor Selection and Regulator Stability**

Capacitors are required at the MAX15101's input and output. Connect at least a 2.2µF capacitor between IN and GND (C<sub>IN</sub>) and a 10µF capacitor between OUT and GND (COLIT). Use only surface-mount ceramic capacitors that have low equivalent series resistance (ESR). Make the input and output traces at least 2.5mm wide, and connect CIN and COUT within 6mm of the IC to minimize the impact of PCB trace inductance.

The output capacitor's ESR can affect stability and output noise. Use capacitors with  $50m\Omega$  or less ESR to ensure stability and optimum transient response. This is particularly true in applications with very low output voltage (< 2V) and high output current (> 0.5A).

Since some capacitor dielectrics can vary over bias voltage and temperature, consult the capacitor manufacturer specifications to ensure that the capacitors meet these requirements over all combinations of operating voltage and temperature.

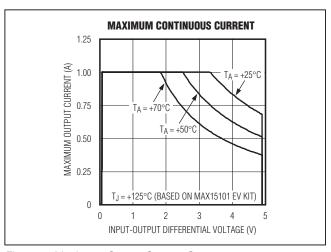


Figure 1. Maximum Output Current Curves

# **Small 1A, Low-Dropout Linear Regulator** in a 2.7mm x 1.6mm Package

#### Input-Output (Dropout) Voltage

A regulator's minimum input-to-output voltage differential (dropout voltage) determines the lowest usable supply voltage. The dropout voltage is simply the product of regulator's RDS(ON) for the pass device and rated load current. The MAX15101's RDS(ON) is  $20m\Omega$ , at 1A this yields a dropout voltage of 20mV (see the Electrical Characteristics and the Dropout Voltage vs. Output Current graph in the *Typical Operating Characteristics*).

#### Noise, PSRR, and Transient Response

The MAX15101 is designed to achieve low dropout voltage while still maintaining good noise, transient response, and AC rejection. When operating from very noisy sources, supply noise rejection and transient response can be improved by increasing the values of the input and output capacitors. The MAX15101 output noise is typically 15µV<sub>RMS</sub> (see the Noise Density graph in the Typical Operating Characteristics).

#### Thermal Considerations in PCB Lavout

The rated package dissipation strongly depends on the copper area connected to the regulator. The maximum power dissipation allowed in the package is 1538mW (measured from the MAX15101 EV kit).

More power dissipation can be handled by the package if great attention is given during PCB layout. For example, using the top and bottom copper as a heatsink and connecting the thermal vias to one of the middle layers (GND) transfers the heat from the package into the board more efficiently, resulting in lower junction temperature at high power dissipation in some MAX15101 applications. Furthermore, the solder mask around the IC area on both top and bottom layers can be removed to radiate the heat directly into the air. The maximum allowable power dissipation in the IC is as follows:

$$P_{MAX} = [(T_{J(MAX)} - T_A)/(\theta_{JA})]$$

where  $(T_{J(MAX)} - T_A)$  is the temperature difference between the maximum-allowed die junction (+150°C) and the surrounding air.

A sample layout is available on the MAX15101 EV kit.

#### **Ordering Information**

	PART	TEMP RANGE	PIN-PACKAGE
MAX	(15101EWL+	-40°C to +85°C	15 WLP

+Denotes a lead(Pb)-free/RoHS-compliant package.

#### **Chip Information**

PROCESS: BICMOS

#### **Package Information**

For the latest package outline information and land patterns (footprints), go to www.maxim-ic.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND
TYPE	CODE	NO.	PATTERN NO.
15 WLP	W151B2+1	<u>21-0205</u>	Refer to Application Note <u>1891</u>

# Small 1A, Low-Dropout Linear Regulator in a 2.7mm x 1.6mm Package

## **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	12/11	Initial release	_

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time. The parametric values (min and max limits) shown in the Electrical Characteristics table are guaranteed. Other parametric values quoted in this data sheet are provided for guidance.