

January 2008

# FMS3110 / FMS3115 Triple Video D/A Converters, 3x10-Bit, 150Ms/s

#### **Features**

- 10-bit Resolution
- 150 Megapixels per Second
- ± 0.1% Linearity Error
- /SYNC and /BLANK Controls
- $1.0V_{PP}$  Video into  $37.5\Omega$  or  $75\Omega$  Load
- Internal Bandgap Voltage Reference
- Double-buffered Data for Low Distortion
- TTL-compatible Inputs
- Low Glitch Energy
- Single +5V Power Supply

## **Applications**

- Video Signal Conversion
  - -RGB
  - -YC<sub>B</sub>C<sub>R</sub>
  - -Composite, Y, C
- Multimedia Systems
- Image Processing
- True-color Graphics Systems: 1 Billion Colors
- **Broadcast Television Equipment**
- High-Definition Television (HDTV) Equipment
- **Direct Digital Synthesis**

## Description

FMS3110 / FMS3115 products are low-cost, triple D/A converters tailored for graphics and video applications where speed is critical. Two speed grades are available:

- FMS3110-100Ms/s
- FMS3115-150Ms/s

TTL-level inputs are converted to analog current outputs that can drive  $25-37.5\Omega$  loads corresponding to doubly terminated  $50-75\Omega$  loads. A sync current following /SYNC input timing is added to the I<sub>OG</sub> output. /BLANK overrides RGB inputs, setting IoG, IoB, and IoR currents to zero when /BLANK = L. Although appropriate for many applications, the internal 1.235V reference voltage can be overridden by the V<sub>REF</sub> input.

Few external components are required: a current reference resistor, current output load resistors, and decoupling capacitors.

Package is a 48-lead LQFP. Fabrication technology is CMOS. Performance is guaranteed from 0 to 70°C.

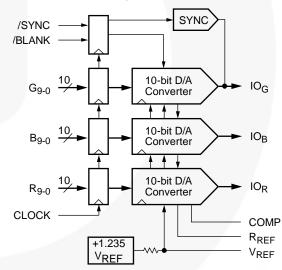


Figure 1. Block Diagram

# **Ordering Information**

Part Numbers	Conversion Rate	Operating Temperature Range	Screening	Package
FMS3110KRC	100Ms/s	0 to 70°C	Commercial	48-Contact Leadless Quad Flat Package
FMS3115KRC	150Ms/s	0 to 70°C	Commercial	48-Contact Leadless Quad Flat Package



All packages are lead free per JEDEC: J-STD-020B standard.

# **Pin Configuration**

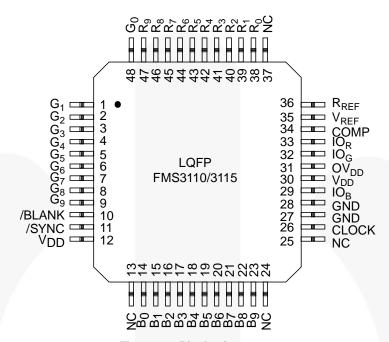


Figure 2. Pin Assignments

## **Pin Definitions**

Name	Pin #	Value	Description
Clock / P	ixel I/O		
CLK	26	TTL	Clock Input. The clock input is TTL-compatible and all pixel data is registered on the rising edge of CLK. It is recommended that CLK be driven by a dedicated TTL buffer to avoid reflection-induced jitter, overshoot, and undershoot.
R9-0	47-37	TTL	<b>Red Pixel Data Inputs</b> . TTL-compatible red data inputs are registered on the rising edge of CLK.
G9-0	48, 9–1	TTL	<b>Green Pixel Data Inputs</b> . TTL-compatible green data inputs are registered on the rising edge of CLK.
B9-0	23–14	TTL	<b>Blue Pixel Data Inputs</b> . TTL-compatible blue data inputs are registered on the rising edge of CLK.
Controls			
/SYNC	11	TTL	Sync Pulse Input. Bringing SYNC LOW turns off a 40 IRE (7.62 mA) current source which forms a sync pulse on the green D/A converter output. /SYNC is registered on the rising edge of CLK with the same pipeline latency as /BLANK and pixel data. /SYNC does not override any other data and should be used only during the blanking interval.
	·		Since this is a single-supply D/A and all signals are positive-going, /SYNC is added to the bottom of the green D/A range. Turning /SYNC OFF means turning the current source ON. When a sync pulse is desired, the current source is turned OFF. If the system does not require sync pulses from the green D/A converter, /SYNC should be connected to GND.
/BLANK	10	TTL	<b>Blanking Input</b> . When /BLANK is LOW, pixel inputs are ignored and the D/A converter outputs fall to the blanking level. /BLANK is registered on the rising edge of CLK and has the same pipeline latency as /SYNC.
Video Ou	tputs		
IO <sub>R</sub>	33	0.714V <sub>pp</sub>	<b>Red Current Output</b> . The current source outputs of the D/A converters are capable of driving RS-343A/SMPTE-170M-compatible levels into doubly terminated $75\Omega$ lines.
IO <sub>G</sub>	32	1V <sub>PP</sub>	<b>Green Current Output</b> . The current source outputs of the D/A converters are capable of driving RS-343A/SMPTE-170M-compatible levels into doubly terminated $75\Omega$ lines. Sync pulses may be added to the green D/A output.
IOB	29	0.714V <sub>PP</sub>	<b>Blue Current Output</b> . The current source outputs of the D/A converters are capable of driving RS-343A/SMPTE-170M-compatible levels into doubly terminated $75\Omega$ lines.
Voltage F	Reference		
$V_{REF}$	35	+1.235V	Voltage Reference Output/Input. An internal voltage source of +1.235V is output on this pin. An external +1.235V reference may be applied here to override the internal reference. Decoupling VREF to GND with a 0.1µF ceramic capacitor is required.
$V_{REF}$	36	560Ω	Current-Setting Resistor. Full-scale output current of each D/A converter is determined by the value of the resistor connected between $R_{REF}$ and GND. Nominal value of $R_{REF}$ is found from: $R_{REF} = 9.1 \; (V_{REF}/I_{FS}) \\$ where $I_{FS}$ is the full-scale (white) output current (in amps) from the D/A converter (without sync). Sync is $0.4 \bullet I_{FS}$ . D/A full-scale (white) current may also be calculated from: $I_{FS} = V_{FS}/R_L \\$ where $V_{FS}$ is the white voltage level and $R_L$ is the total resistive load (in $\Omega$ ) on each D/A converter. $V_{FS}$ is the blank to full-scale voltage.
COMP	34	0.1µF	Compensation Capacitor. A $0.1\mu F$ ceramic capacitor must be connected between COMP and $V_{DD}$ to stabilize internal bias circuitry.
Power ar	d Ground	•	
$V_{DD}$	12, 30, 31	+5V	Power Supply.
GND	27, 28	0.0V	Ground.

## **Absolute Maximum Ratings**

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only. See Figure 4.

Symbol	Parameter	Min.	Max.	Unit
Power Supply V	oltage	•	•	
$V_{DD}$	Measured to Ground	-0.5	7.0	V
Inputs	·			
$V_{IN\_A}$	Applied Voltage, Measured to Ground <sup>(1)</sup>	-0.5	V <sub>DD</sub> +0.5	V
I <sub>IN_F</sub>	Forced Current <sup>(2, 3)</sup>	-10	+10	mA
Outputs				
$V_{OUT\_A}$	Applied Voltage, Measured to Ground <sup>(1)</sup>	-0.5	V <sub>DD</sub> +0.5	V
I <sub>OUT_F</sub>	Forced Current <sup>(2, 3)</sup>	-60	+60	mA
t <sub>SC</sub>	Short-Circuit Duration, Single Output in HIGH State to Ground		Infinite	s
Temperature				
T <sub>A</sub>	Operating Ambient Temperature	-20	+110	°C
$T_J$	Junction Temperature		+150	°C
T <sub>L</sub>	Lead Soldering, 10 Seconds		+300	°C
$T_VP$	Vapor Phase Soldering, 1 Minute		+220	°C
T <sub>STG</sub>	Storage Temperature	-65	+150	°C

#### Notes:

- 1. Applied voltage must be current limited to specified range.
- 2. Forcing voltage must be limited to specified range.
- 3. Current is specified as conventional current flowing into the device.

# **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Paramete	er	Min.	Nom.	Max.	Unit
$V_{DD}$	Power Supply Voltage		4.75	5.00	5.25	V
f	Conversion Rate	FMS3110		7	100	Mono
f <sub>S</sub>	Conversion Rate	FMS3115			150	Msps
	CLK Dulas width LICH	FMS3110	3.1			200
t <sub>PWH</sub>	CLK Pulse-width, HIGH	FMS3115	2.5		У	ns
	CLK Dulas width 1 OW	FMS3110	3.1			
t <sub>PWL</sub>	CLK Pulse-width, LOW		2.5			ns
_	CLK Dulas width	FMS3110	10			
t <sub>W</sub>	CLK Pulse-width	FMS3115	6.6			ns
t <sub>S</sub>	Input Data Setup Time		1.7			ns
t <sub>h</sub>	Input Data Hold Time		0			ns
V <sub>REF</sub>	Reference Voltage, Exter	nal	1.000	1.235	1.500	V
C <sub>C</sub>	Compensation Capacitor			0.1		μF
R <sub>LOAD</sub>	Output Load			37.5		Ω
V <sub>IH</sub>	Input Voltage, Logic HIGH		2.0		$V_{DD}$	V
V <sub>IL</sub>	Input Voltage, Logic LOV	GND		0.8	V	
T <sub>A</sub>	Ambient Temperature, St	ill Air	0		70	°C

## **Electrical Characteristics**

Symbol	Parameter	Conditions <sup>(6)</sup>	Min.	Typ. <sup>(4)</sup>	Max.	Units
I <sub>DD</sub>	Power Supply Current <sup>(5)</sup>	V <sub>DD</sub> =Max.			125	mA
P <sub>D</sub>	Total Power Dissipation <sup>(5)</sup>	V <sub>DD</sub> =Max.			655	mW
Ro	Output Resistance			100		kΩ
Co	Output Capacitance	I <sub>OUT</sub> =0mA			30	рF
l <sub>IH</sub>	Input Current HIGH	V <sub>DD</sub> =Max., V <sub>IN</sub> =2.4V			-5	μA
I <sub>IL</sub>	Input Current LOW	V <sub>DD</sub> =Max., V <sub>IN</sub> =0.4			+5	μΑ
I <sub>REF</sub>	V <sub>REF</sub> Input Bias Current			0	±100	μΑ
$V_{REF}$	Reference Voltage Output			1.235		V
V <sub>oc</sub>	Output Compliance	Referred to V <sub>DD</sub>	-0.4	0	+1.5	V
C <sub>DI</sub>	Digital Input Capacitance			4	10	pF

#### Notes:

- 4. Values shown are typical for V<sub>DD</sub>=+5V and T<sub>A</sub>=25°C.
- 5. Minimum and Maximum values with  $V_{DD}$ =Max and  $T_A$ =Min.
- 6.  $V_{REF}=1.235V$ ,  $R_{LOAD}=37.5\Omega$ ,  $R_{REF}=540\Omega$ .

# **Switching Characteristics**

Symbol	Parameter	Conditions <sup>(8)</sup>	Min.	Typ. <sup>(7)</sup>	Max.	Units
$t_D$	Clock to Output Delay	V <sub>DD</sub> =Min.		10	15	ns
tskew	Output Skew			1	2	ns
t <sub>R</sub>	Output Rise Time	10% to 90% of Full Scale			3	ns
t <sub>F</sub>	Output Fall Time	90% to 10% of Full Scale			3	ns

### Notes:

- 7. Values shown are typical for  $V_{DD}=+5V$  and  $T_A=25$ °C.
- 8.  $V_{REF}=1.235V$ ,  $R_{LOAD}=37.5\Omega$ ,  $R_{REF}=590\Omega$ .

# **System Performance Characteristics See Figure 3**

Symbol	Parameter	Conditions <sup>(10)</sup>	Min.	Typ. <sup>(9)</sup>	Max.	Units
ELI	Integral Linearity Error	V <sub>DD</sub> , V <sub>REF</sub> =Nominal		±0.10	±0.25	%/FS
ELD	Differential Linearity Error	V <sub>DD</sub> , V <sub>REF</sub> =Nominal		±0.10	±0.25	%/FS
EDM	DAC to DAC Matching	V <sub>DD</sub> , V <sub>REF</sub> =Nominal		3	10	%
PSRR	Power Supply Rejection Ratio				0.05	%/%

### Notes:

- 9. Values shown are typical for  $V_{DD}$ =+5V and  $T_A$ =25°C.
- 10.  $V_{REF}$ =1.235V,  $R_{LOAD}$ =37.5 $\Omega$ ,  $R_{REF}$ =590 $\Omega$ .

# **Timing Information**

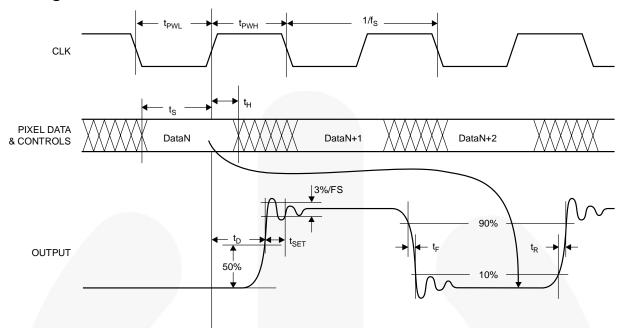
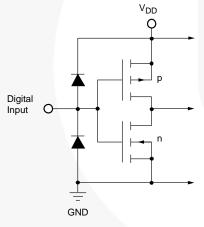


Figure 3. Timing Diagram

# **Equivalent Circuits**



V<sub>DD</sub> O OUT

Figure 4. Equivalent Digital Input Circuit

Figure 5. Equivalent Analog Output Circuit

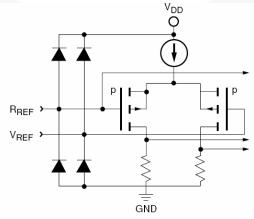


Figure 6. Equivalent Analog Input Circuit

## **Functional Description**

Within the FMS3110/3115 are three identical 10-bit D/A converters, each with a current source output. External loads are required to convert the current to voltage outputs. Data inputs RGB7-0 are overridden by the /BLANK input. /SYNC = H activates sync current from  $l_{OS}$  for sync-on-green video signals.

### **Digital Inputs**

All digital inputs are TTL-compatible. Data is registered on the rising edge of the CLK signal. Following one stage of pipeline delay, the analog output changes  $t_{DO}$  after the rising edge of CLK.

#### /SYNC and /BLANK

/SYNC and /BLANK inputs control the output level (Figure 7 and Table 1) of the D/A converters during CRT retrace intervals. /BLANK forces the D/A outputs to the blanking level, while /SYNC = L turns off a current source connected to the green D/A converter. /SYNC = H adds a 40 IRE sync pulse to the green output; /SYNC = L sets the green output to 0.0V during the sync tip. /SYNC and /BLANK are registered on the rising edge of CLK.

/BLANK gates the D/A inputs and sets the pedestal voltage. If /BLANK = HIGH, D/A inputs are added to a pedestal, which offsets the current output. If /BLANK = LOW, data inputs and the pedestal are disabled.

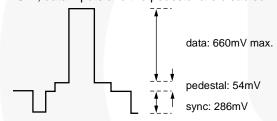


Figure 7. Nominal Output Levels

### **D/A Outputs**

Each D/A output is a current source. To obtain a voltage output, a resistor must be to ground. Output voltage depends upon this external resistor, the reference voltage, and the value of the gain-setting resistor connected between  $R_{\text{REF}}$  and GND.

Normally, a  $75\Omega$  source termination resistor is connected between the D/A current output pin and GND near the D/A converter. A  $75\Omega$  line can be connected with another  $75\Omega$  termination resistor at the far end of the cable. This "double termination" presents the D/A converter a net resistive load of  $37.5\Omega$ .

The FMS3110/3115 may also be operated with a single  $75\Omega$  terminating resistor. To lower the output voltage swing to the desired range, the nominal value of the resistor on R<sub>REF</sub> should be doubled.

### **Voltage Reference**

All three D/A converters are supplied with a common voltage reference. Internal bandgap voltage reference voltage is +1.235V with a  $3K\Omega$  source resistance. An external voltage reference may be connected to the  $V_{REF}$  pin, overriding the internal voltage reference.

A  $0.1\mu F$  capacitor must be connected between the COMP pin and  $V_{DD}$  to stabilize internal bias circuitry and ensure low-noise operation.

#### **Power and Ground**

Required power is a single +5.0V supply. To minimize power-supply induced noise, analog +5V should be connected to  $V_{DD}$  pins with 0.1 and 0.01 $\mu$ F decoupling capacitors placed adjacent to each  $V_{DD}$  pin or pin pair.

The high slew-rate of digital data makes capacitive coupling to the outputs of any D/A converter a potential problem. Since the digital signals contain high-frequency components of the CLK signal, as well as the video output signal, the resulting data feed-through often looks like harmonic distortion or reduced signal-to-noise performance. All ground pins should be connected to a common solid ground plane for best performance.

Table 1. Output Voltage vs. Input Code, /SYNC and /BLANK (V<sub>REF</sub>=1.235V, R<sub>REF</sub>=590Ω, R<sub>L</sub>=37.5Ω)

DCD (MCD LCD)	Blu	e and Red D	/As	Green D/A			
RGB <sub>9-0</sub> (MSBLSB)	/SYNC	/BLANK	V <sub>out</sub>	/SYNC	/BLANK	V <sub>out</sub>	
11 1111 1111	X	1	0.7140	1	1	1.0000	
11 1111 1111	Х	1	0.7140	0	1	0.7140	
11 1111 1110	X	1	0.7134	1	1	0.9994	
11 1111 1101	X	1	0.7127	1	1	0.9987	
:	:	:	:	:	:	:	
10 0000 0000	Х	1	0.3843	1	1	0.6703	
01 1111 1111	X	1	0.3837	1	1	0.6697	
:	:	:	:	:	:	:	
00 0000 0010	Х	1	0.0553	1	1	0.3413	
00 0000 0001	Х	1	0.0546	1	1	0.3406	
00 0000 0000	Х	1	0.0540	1	1	0.3400	
XX XXXX XXXX	Х	0	0.0000	1	0	0.2860	
XX XXXX XXXX	Χ	0	0.0000	0	0	0.0000	

## **Application Information**

Figure 8 illustrates a typical FMS3110/3115 interface circuit. In this example, an optional 1.2V bandgap reference is connected to the  $V_{\text{REF}}$  output, overriding the internal voltage reference source.

### Grounding

It is important that the FMS3110/3115 power supply be well-regulated and free of high-frequency noise. Careful power supply decoupling ensures the highest quality video signals at the output of the circuit. The FMS3110/3115 has separate analog and digital circuits. To keep digital system noise from the D/A converter, it is recommended that power supply voltages ( $V_{\rm DD}$ ) come from the system analog power source and all ground connections (GND) be made to the analog ground plane. Power supply pins should be individually decoupled at the pin.

## **Printed Circuit Board Layout**

Designing with high-performance mixed-signal circuits demands printed circuits with ground planes. Overall system performance is strongly influenced by the board layout. Capacitive coupling from digital to analog circuits may result in poor D/A conversion. Consider the following suggestions when designing the layout:

 Keep the critical analog traces (V<sub>REF</sub>, I<sub>REF</sub>, COMP, I<sub>OS</sub>, I<sub>OR</sub>, I<sub>OG</sub>) as short as possible and as far as possible from all digital signals. The FMS3110/3115 should be located near the board edge, close to the analog output connectors.

- 2. The power plane for the FMS3110/3115 should be separate from that which supplies the digital circuitry. A single power plane should be used for all of the V<sub>DD</sub> pins. If the power supply for the FMS3110/3115 is the same as that of the system's digital circuitry, power to the FMS3110/3115 should be decoupled with 0.1μF and 0.01μF capacitors and isolated with a ferrite bead.
- The ground plane should be solid, not crosshatched. Connections to the ground plane should have very short leads.
- 4. If the digital power supply has a dedicated power plane layer, it should not be placed under the FMS3110/3115, the voltage reference, or the analog outputs. Capacitive coupling of digital power supply noise from this layer to the FMS3110/3115 and its related analog circuitry can have an adverse effect on performance.
- CLK should be handled carefully. Jitter and noise on this clock degrade performance. Terminate the clock line carefully to eliminate overshoot and ringing.

#### **Related Products**

- FMS38XX Triple 8-bit 150Msp D/A Converters
- FMS9884A 3x8-bit 140Ms/s A/D Converter

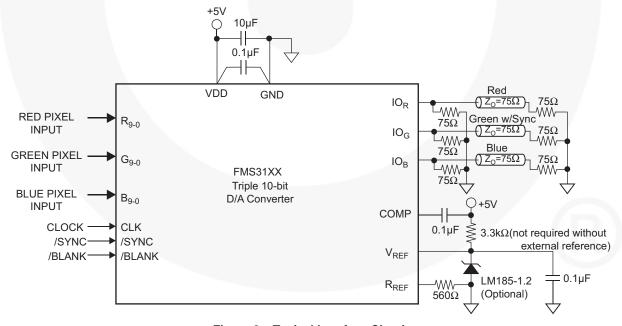


Figure 8. Typical Interface Circuit

## **Physical Dimensions**

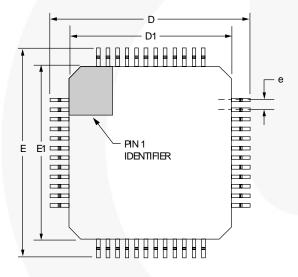
Ombol	Inc	hes	Millim	neters	Notes	
Symbol	Min.	Max.	Min.	Max.	Notes	
Α	.055	.063	1.40	1.60		
A1	.001	.005	.05	.15	- 4/	
A2	.053	.057	1.35	1.45		
В	.006	.010	.17	.27	7	
D/E	.346	.362	8.8	9.2	8	
D1/E1	.268	.284	6.8	7.2	2	
е	.019	BSC	.50			
L	.017	.029	.45	.75	6	
N	48		48		4	
ND	12		12		5	
α	0	7	0	7		
ccc	.0	04	0.0	08		

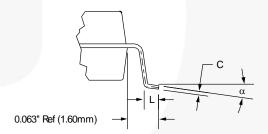
#### Notes:

- 1. All dimensions and tolerances conform to ANSI Y14.5M-1982.
- Dimensions "D1" and "E1" do not include mold protrusion. Allowable protrusion is 0.25mm per side. D1 and E1 are maximum plastic body size dimensions including mold mismatch.
- 3. Pin 1 identifier is optional.
- 4. Dimension ND: Number of terminals.
- 5. Dimension ND: Number of terminals per package edge.
- 6. "L" is the length of terminal for soldering to a substrate.
- 7. Dimension "B" does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum B dimension by more than 0.08mm. Dambar can not be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07mm for 0.4mm and 0.5mm pitch packages.

To be determined at seating place NCN

8.





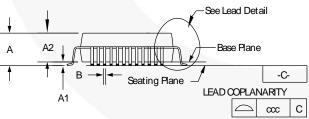


Figure 9. 48-Contact Leadless Quad Flat Package (LMQFP)

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QS™

QT Optoelectronics™ Quiet Series™ RapidConfigure™

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  - device, or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness

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